

Semiconducting SWCNT Photo Detector for High Speed Switching Through Single Halo Doping

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Abstract: The method opted for accuracy, and no existing studies are based on this method. A design and characteristic survey of a new small band gap semiconducting Single Wall Carbon Nano Tube (SWCNT) Field Effect Transistor as a photodetector is carried out. In the proposed device, better performance is achieved by increasing the diameter and introducing a new single halo (SH) doping in the channel length of the CNTFET device. This paper is a study and analysis of the performance of a Carbon Nano Tube Field Effect Transistor (CNTFET) as a photodetector using the self-consistent Poisson and Green function method. The 2D self-consistent Poisson and Green's function method for various optical intensities and wavelength simulate this proposed photodetector. The performance study is based on the simulation of drain current, transconductance, sub-threshold swing, cut-off frequency, gain, directivity, and quantum efficiency under dark and illuminated conditions. These quantum simulation results show that cut-off frequency increases while there is an increase in diameter. The proposed SH-CNTFET provides better performance in terms of higher gain and directivity than conventional CNTFET (C-CNTFET). This device will be helpful in optoelectronic integrated circuits (OEIC) receivers due to its superior performance.

Keywords: Single halo CNTFET; self-consistent poisson equation; green's function; photodetector

1 Introduction

CNTFETs have been studied, and their characteristics have been analyzed in various forms. CNTFETs have been designed and fabricated with varying parameters, and evolution is still in the making. Significant studies related to the proposed model include the following. The Gaussian-doping distribution in the channel region of the tunnel CNTFET structure exhibits increased saturation current but decreased OFF-state current [1]. The implantation of halo doping at the source side and a lightly doped drain enhances the performance of



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the Silicon Nanotube Tunnel Field Effect Transistor (NT TFET) [2]. Simulation has been carried out, and experimental data of single and multi-tube carbon nanotube field effect-transistors (CNTFETs) have been acquired. Device characteristics with different channel lengths and Schottky barrier heights have been discussed [3]. The effect of halo doping on the thermal noise behaviour of halo-doped MOSFETs and the drain current power spectral density (SID) of the thermal noise of halo MOSFETs are analyzed [4]. The electrical characteristics of CNTFET with dielectric materials such as HfO_2 , SiO_2 and ZrO_2 were assessed. FET toy, a Nano device simulator, was used for this purpose [5]. It has been found that the Double-Halo-Doping CNTFET is a potential candidate in the study to establish the maintenance of threshold voltage and drain-induced barrier lowering (DIBL) [6]. The two-dimensional non-equilibrium Green function, solved self-consistently with Poisson's equation of the Schottky barrier, establishes that the carbon nanotube field-effect transistor (SB-CNTFET) structures decrease the leakage current [7]. The behaviour of capacitances in halo channel MOSFET has been investigated for various cases of non-uniformity doping using TCAD simulations [8]. Four different electrodes were used in the single-walled CNT, and the photoconductivity was experimentally measured [9]. Graphene Nano Ribbon Field Effect Transistors (GNRFET) with halo doping are simulated to obtain the highest ratio of $I_{\text{ON}}/I_{\text{OFF}}$ [10]. Graphene Nano Ribbon field-effect transistors (GNRFETs) with a single halo pocket in the intrinsic channel region do weaken band-to-band tunnelling (BTBT) [11]. A study of the high-frequency characteristics of Single Halo-doped Carbon Nanotube Field-Effect Transistors (SH-CNTFETs) has been done [12]. The Carbon Nanotube and the Density of State (DOS) with different types of nanotubes considering the energy gap have been reviewed [13]. The halo and linear doping CNTFET (HL-CNTFET) structure decreases the drain leakage current and improves high frequency [14]. It is established that the carrier efficiency and the Short-Channel Effects (SCE) are improved, and the same has been done through a quantum simulation of a single halo dual-material gate CNTFET [15]. The cut-off frequency of single halo implantation in a channel of a lightly doped drain and the source is investigated [16]. The carbon nanotube design is functionalized with suitable molecules, and the transmission spectrum and I-V characteristics are studied [17]. CNTs with small band-gap chirality are designed and are found to have faster switching CNTFETs [18]. Band-to-band tunnelling is well suppressed in the symmetric graded double halo channel (GDH-CNTFET) [19]. A decrease in the width of the depletion region with a resultant increase in device current is observed in the implementation of pocket implantation, which is halo-doped and placed at the source end [20]. Under illuminated conditions, the efficiency due to monochromatic light and the photovoltaic effect is calculated [21]. The self-consistent solution of the 2D Poisson and Schrodinger equations was used for the quantum simulation of CNTFET. This simulation improved the transconductance characteristics and drain conductance by varying the source and drain parameters [22]. Device performance has been investigated, focusing on transconductance characteristics, output characteristics with different gate oxide thicknesses, and carbon nanotube diameter [23]. The increase of the length and HD region concentration with halo doping in a CNTFET improved the ON-OFF current ratio, the leakage current and the subthreshold swing [24]. The ambipolar behaviour of CNTFET is found to enhance the performance of subthreshold swing. A reduced drain-induced barrier is observed with lower symmetric double-halo (DH)-doping in CNTFETs [25]. A significant decrease in leakage current and drain conductance is observed from the quantum simulation of Single Halo CNTFET [26]. The analogue/RF performance of silicon nanotube FET can be improved with halo doping at the drain end. A decrease in threshold voltage increases tube diameter, and this property is suitable for radio-frequency applications [27]. Silicon Nanotube Tunnel Field Effect Transistor (NT-TFET) improves the subthreshold swing and short channel effects for ultra-low power applications [28].

CNT has high strength, durability, electrical conductivity, thermal conductivity, and lightweight properties compared to conventional semiconductor materials and carries a high current. Due to their high

conductivity, CNTs are used in electronic devices. The proposed device has the following modifications which achieve improved performance: Firstly, the design of a CNTFET photodetector, which is Single Halo-Doped and has a single gate material, A_u , to achieve quantum efficiency, and has a control on the length of the channel and directivity. The single halo doping is applied at the drain end. The size of doping L_{shd} is increased to 5 nm. The doping concentration (N_{shd}) is set as 1.2 nm^{-1} . These settings help to achieve reduced leakage current. Secondly, the implementation of the zigzag CNT ($n=25$, $m=0$) having a diameter of 1.95 nm, which is larger than the existing models, and an energy band gap of 0.4401 eV which is smaller than the current models, results in increased cut-off frequency. The Third significant modification is the HfO_2 gate oxide insulator sandwiching the CNTFET. The thickness of the gate oxide insulator t_{ox} is 1.4 nm, and the dielectric constant k is 30. This modification increases the transconductance of the proposed SH-CNTFET.

This article studies and analyses the performance of a Carbon Nano Tube Field Effect Transistor (CNTFET) as a photodetector using the self-consistent Poisson and Green function method. The proposed modelling and simulation of the single wall Single Halo Carbon Nano Tube Field Effect Transistor (SH-CNTFET) has been divided into five sections. The introduction and previous literature are discussed in Section 1. In Section 2, the modelling of the proposed device is discussed. Simulation of the proposed SH-CNTFET using the 2D Poisson equation and the Green function is discussed in Section 3. Section 4 is the result and discussion of the proposed SH-CNTFET, and Section 5 is the conclusion process. The existing CNTFET sets back in terms of transconductance, gain, directivity etc. To overcome this concern, the proposed SH-CNTFET is proposed to achieve a better performance in the parameters above.

2 Modeling of Single Halo CNTFET Structure

The proposed Single Halo CNTFET is designed as in Fig. 1. The parameters used in the simulation of the proposed structure have been listed in Table 1. Numerical computation has been carried out for the SH-CNTFET. The diagrams were plotted using Matlab.

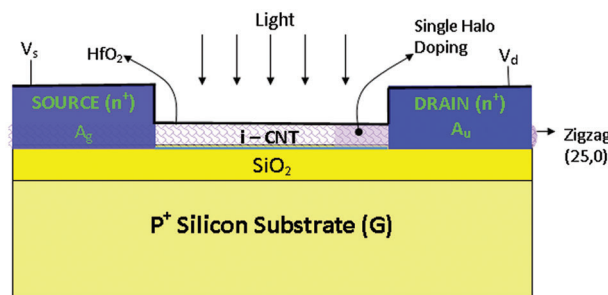


Figure 1: Cross-sectional view of an SH-CNTFET

In the proposed planar SH-CNTFET structure, zigzag ($n=25$, $m=0$) chirality semiconductor type CNT is used, and the diameter of CNT (d_{cnt}) is 1.95 nm. The energy band gap is 0.4401 eV, and the CNT bridges two electrodes. The layers are the source (A_g) and drain (A_u) and are coupled with a thin layer of HfO_2 , with an insulating oxide thickness (t_{ox}) of 1.4 nm and a high dielectric constant (k) of 30. The device dimensions for the structures include a gate (back-gated) length (L_g) of 15 nm and an undoped channel length (L_c) of 10 nm. The source length (L_s) and drain length (L_d) are 20 nm each. Silver metal ($A_g=4.2-4.5$ eV) is used at the source, and gold metal ($A_u=5.3$ eV) is used at the drain. The $n+$ source and drain doping density are $1.5 \times 10^7 \text{ cm}^{-1}$. The carbon atom density of ($4 n/3 \text{ acc}$) in CNTFET, which is 100 nm^{-1} , is compared with the above doping concentration. The length of the undoped or intrinsic CNT (i-CNT) channel is 10 nm. The channel consists of a single halo p-doped CNT region with a length of L_{shd} . The

halo doping is done at the drain end. The length of the halo-doped region on the drain side of the Single Halo-Carbon Nano Tube Field Effect Transistor (SH-CNTFET) is $L_{\text{shd}} = 5$ nm, and the doping concentration is $N_{\text{shd}} = 1.2 \text{ nm}^{-1}$. Except for halo doping, the values of the Single Halo CNTFET and C-CNTFET Device Parameters are equal. V_g is a gate-controlled voltage, while V_d and V_s are the drain and source voltages applied to the SH-CNTFET. The prime focus of the proposed SH-CNTFET simulation is on the characteristic study in dark and illuminated conditions with light incident power $P_{\text{opt}} = 100 \text{ W/cm}^2$ and photon energy of 2.72 eV. Wavelength (λ) = 456 nm.

Table 1: Parameter settings

Parameter	Value
Length of the gate (L_g)	15 nm
Length of single halo (L_{shd})	5 nm
Length of the source (L_s)	20 nm
Length of the drain (L_d)	20 nm
Length of the i-channel (L_c)	10 nm
Zigzag (n, m)	(n = 25, m = 0)
Diameter of the CNT	1.95 nm
Single halo doping concentration (N_{shd})	1.2 nm^{-1}
Source doping concentration (N_s)	$1.5 \text{ cm} \times 10^7 \text{ cm}^{-1}$
Drain doping concentration (N_d)	$1.5 \text{ cm} \times 10^7 \text{ cm}^{-1}$
Gate material work function (ϕ_g)	5.3 eV
Gate oxide thickness (t_{ox})	1.4 nm
High-K insulator (HfO_2)	30

3 Quantum Simulation of SH-CNTFET

The proposed model is implantation, which has already established that SCEs can be reduced and achieved in a single-halo CNTFET. Hence, the SH-CNTFET photodetector has been modelled using a two-dimensional (2-D) quantum simulation. The gain, output and transconductance characteristics, cut-off frequency, responsivity and directivity of SH-CNTFET have been obtained by solving, under open boundary conditions, the Schrodinger and 2-Dimensional Poisson equations using self-consistent methods. This solution has been received within Green's function (NEGF) framework. In this simulation of SH-CNTFET done under room temperature, below are the locations of the Fermi levels. The valence band A_g is set at the top, and the Fermi level A_u is set for the conduction band. At the junction between the metal and SH-CNTFET is the electron-electron coupling and the photon-electron coupling, described in Green's function equations with their self-energies. The excitonic effect can be neglected in the given simulation. Even though uneven metal contacts result in doping, this neglect is reasonable. The I-V characteristics are obtained by the self-consistent solution of the Poisson equations and Green function. The simulation process is detailed in the steps below. The Green's function for the given device is given by

$$G(E) = [(E + i\eta^+ - qU)I - H - \varepsilon_s - \varepsilon_d - \varepsilon_{\text{pho}}]^{-1} \quad (1)$$

In Eq. (1), E denotes energy, and infinitesimal broadening is denoted by η . The identity matrix is given as I. Other parameters include q, which is the electron charge, and U, which is the electrostatic potential. The self-energy at the drain and source ends of the metal contacts is given individually ε_s , and ε_d calculated. H

shows the Hamiltonian matrix of the CNT. A tight-binding Hamiltonian describes the SH-CNTFET, where the Hamiltonian is organized with one orbital per grid. Considering thickness and given that m^* is the effective mass of the electron, the hopping parameter of the nearest neighbour in the above-described SH-CNTFET is provided by $t_0 = \hbar^2/2m^*a^2$. The double bond length between the carbon atoms contains the effective mass. The open boundary condition of the SH-CNTFET is given by coupling the metal contacts with the SH-CNTFET.

The Hamiltonian matrix can be calculated as

$$H = \begin{bmatrix} U_1 & b_{2q_1} & \dots & \dots & \dots & \dots \\ b_{2q_1} & U_2 & t & \dots & \dots & \dots \\ \dots & b_{2q_1} & U_3 & \dots & \dots & \dots \\ \vdots & \vdots & \vdots & \vdots & \vdots & \vdots \\ \vdots & \vdots & \vdots & \vdots & \vdots & t \\ \dots & \dots & \dots & \dots & t & U_N \end{bmatrix} \quad (2)$$

where $b_{2q} = t \cos(q/n)$, $t = 3$ eV, n is the chirality vector of the nanotube, q is the angular quantum number, N is the total number of carbon rings along the device, and t is the potential for the interaction of two adjacent positions in the network. The total number of carbon atoms is N . The size of the Hamiltonian matrix is $N \times N$.

The self-energy ε_{pho} is generated by the electron-photon coupling. This is expressed as

$$\varepsilon_{pho} = -\left(\frac{i}{2}\right)\Gamma_{pho} \quad (3)$$

The electron-photon coupling generates the broadening function, which is given by Γ_{ph} .

$$\Gamma_{phO} = \varepsilon_{phO}^{in}(E) + \varepsilon_{phO}^{out}(E) \quad (4)$$

ε_{phO}^{in} and ε_{phO}^{out} are in-and-out scattering functions for contact coupling, which are defined by [21]

$$\varepsilon_{phO}^{in}(E) = N\alpha \left[\sum_{kj} M_{rk} M_{js} \left(G_{jj}^n(E + h\mu) + G_{kj}^n(E - h\mu) \right) \right] + \alpha \sum_{kj} M_{rk} M_{js} G_{jj}^n(E + h\mu) \quad (5)$$

$$\varepsilon_{phO}^{out}(E) = N\alpha \left[\sum_{kj} M_{rk} M_{js} \left(G_{kj}^p(E - h\mu) + G_{kj}^p(E + h\mu) \right) \right] + \alpha \sum_{kj} M_{rk} M_{js} G_{kj}^p(E - h\mu) \quad (6)$$

In the above equation, j is the layer count of the grid slice. $M_{lm} = \pm \delta_{l\pm 1, m}$. In addition, $\alpha = q^2 a^2 \gamma^2 F / (2h\mu c \varepsilon_0)$ given that q is the electron charge and the measure of the thickness of the grid a . Other parameters include the coupling strength of subsequent carbon atoms, which is given by, γ and the light speed, given by c , the permittivity value of the vacuum is given by ε_0 , and the energy of the light is given by $h\mu$.

The Correlation function of the hole and electron in the Eq. (4) can be written as

$$G^n(E) = G(E) \left[\Gamma_s(E) f_s(E) + \Gamma_D(E) f_D(E) + \varepsilon_{pho}^{in}(E) \right] G^+(E) \quad (7)$$

$$G^p(E) = G(E) \left\{ \Gamma_s(E) [1 - f_s(E)] + \Gamma_D(E) [1 - f_D(E)] + \varepsilon_{pho}^{out}(E) \right\} G^+(E) \quad (8)$$

where $f_{S,D}$ is the Fermi–Dirac function of the source and drain metal contacts, and the broadening function is $\Gamma_{S,D} = i(\varepsilon_{S,D} - \varepsilon_{S,D}^+)$. It is obtained by the contact the source and drains at the terminal.

The back gate SH-CNTFET's self-consistent electrostatic potential (U) in Eq. (1) can be calculated using a two-dimensional Poisson equation. The value of the potential for the Poisson equation can be given as

$$\frac{\partial U(x, y)}{\partial y^2} + \left(\frac{1}{y} + \frac{1}{\varepsilon} \frac{\partial \varepsilon}{\partial y} \right) \frac{\partial U(x, y)}{\partial y} + \frac{\partial U(x, y)}{\partial x^2} = - \frac{qQ}{\varepsilon} \quad (9)$$

where ε is the permittivity, q is the electron charge, and the CNT is along the x -axis, the y -axis is perpendicular to the planar SH-CNTFET, U is the electrostatic surface potential of the CNT, and Q is the charge density of the CNT.

$$Q = N_d^+ - N_a^- + \frac{4}{\nabla z} \int_{-\infty}^{E_{m(j)}} \frac{G_{jj}^p(E)}{2\pi} dE - \frac{4}{\nabla z} \int_{E_{m(j)}}^{+\infty} \frac{G_{jj}^n(E)}{2\pi} dE \quad (10)$$

where N_d^+ , N_a^- are the donor, and acceptor concentrations of carbon nanotube, respectively; j is the layer count on the grid slice, and $E_m(j)$ is the mid-gap energy. The assumption is that the charge density Q is uniformly distributed on the CNT surface as depicted in Fig. 1a.

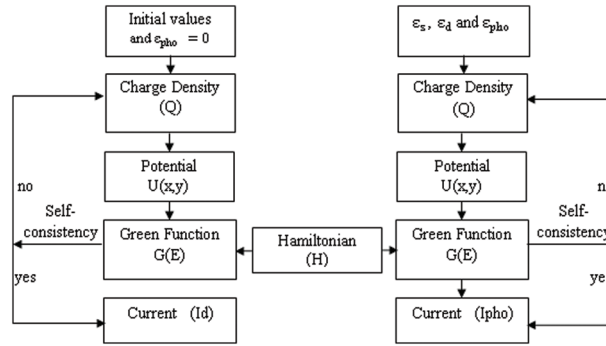


Figure 1a: The process of current calculation for dark and illuminated conditions

In the simulation of self-consistency, the dark current (I_d) is calculated using the following steps: In the first step, assign the value $\varepsilon_{ph0} = 0$ in the equation to the Green function at Eq. (1). Then, using the born loop between Eqs. (7) and (10) with, calculate Q and U with $\varepsilon_{ph0} = 0$. In addition, at the next step, the electrostatic potential U is calculated using the finite difference method from Eq. (9) with $Q=0$ on the right side of the Poisson equation. In the further step, once self-consistency is achieved, the dark $G(E)$ is obtained using the green function method from Eq. (1). In the following step, $G_n(E)$ and $G_p(E)$ are calculated in the dark condition using Eqs. (7) and (8).

The current in dark conditions is calculated by

$$I = \left(\frac{4q}{2\pi h} \right) \int_{-\infty}^{+\infty} T_r \left[H_{j,j+1} G_{j,j+1}^n(E) - H_{j+1,j} G_{j+1,j}^n(E) \right] dE \quad (11)$$

where h is the Plank's constant, q is the electron charge, T_r is the transmission coefficient of electron tunneling through the channel [12].

Self-consistent photocurrent can be calculated using the following steps. At the start, the initial values are derived from Eqs. (5) and (6). The values in the dark conditions for $G_n(E)$ and $G_p(E)$ are used. The self-energy due to electron-photon coupling ε_{ph0} and $G(E)$ are then solved iteratively using Eqs. (1), (3) and (8). $G_n(E)$ and $G_p(E)$ are calculated under illuminated conditions. Photocurrent (I_{ph}) is calculated using Eq. (11). The I-V curve is generated in illuminated and dark conditions by repetition of the above procedure with

varying bias voltages. The intrinsic cut-off frequency of the transistor is calculated using $f_{cut} = \frac{1}{2\pi} \frac{g_m}{c_g}$ where g_m is transconductance, c_g is the gate capacitance. Where $g_m = \frac{\partial I_d}{\partial V_g}$ and $c_g = \frac{\partial Q_g}{\partial V_g}$ where Q_g is total charge on the gate. The gain of the SH-CNTFET is $A_v = \frac{g_m}{g_d}$. The drain conductance is g_d . More details of our simulation have been investigated [12]. The Responsivity of the SH-CNTFET is calculated as $R = \frac{I_{ph}}{P_{opt}}$. The photocurrent obtained under illuminated conditions is I_{ph} . P_{opt} is the incident light on the device, which is also the optical power intensity. The quantum efficiency can be expressed in terms of photon energy and electron charge as $\eta = \frac{I_{ph}/q}{P_{opt}/h\nu} \times 100$. Here $h\nu$ is the photon energy, and q is the electron charge. The directivity of the SH-CNTFET can be calculated as $D = \sqrt{\frac{A}{2qI_d}} R$ where A is the area of the device, q is the electron charge and, I_d is the dark current of the SH-CNTFET, and as has been discussed in [9].

4 Result and Discussion

Fig. 2 shows the simulation results of the current drain characteristics of the zigzag (25, 0) semiconductor CNT. Where $V_{gs} = 1.2$ V, $P_{opt} = 100$ W/cm², photon energy is 2.72 eV. $t_{ox} = 1.4$ nm, and wavelength (λ) = 456 nm. At V_{ds} 0.8 V, the current increases linearly and reaches a saturated state. It is observed that the I_{ph} of SH-CNTFET is larger than the dark current (I_d) of SH-CNTFET. The latter is lower when the I_{off} of C-CNTFET and the SH-CNTFET are compared. Thus, the drain current of the Single Halo CNTFET is higher than C-CNTFET. The leakage current decreases in SH-CNTFET, and the on-off current ratio increases when compared to C-CNTFET [25]. Thus, the SH-CNTFET structure can do the following. The ON/OFF current ratio is increased. The structure demonstrates that the gate controls the channel and, finally and more importantly, improves the switching speed. The proposed device has a high switching speed, which makes it ideal for high-speed and high-frequency circuits [12]. The drop in conductance at the drain for SH-CNTFET and a rise in transconductance are obtained in the case of SH-CNTFET. Then the transconductance of the SH-CNTFET is 7.8×10^{-5} mho.

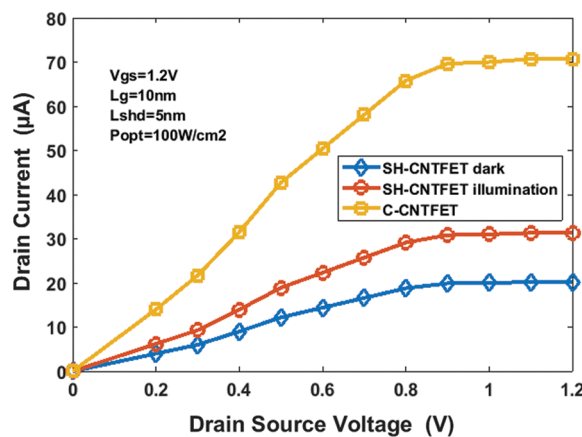


Figure 2: Drain current characteristic at $V_{gs} = 1.2$ V, $P_{opt} = 100$ W/cm², wavelength (λ) = 456 nm under dark and illumination conditions

SH-CNTFET and C-CNTFET do not coincide. However, with the single halo-doped at the drain side, the dark current of SH-CNTFET is significantly decreased. It proves that the lowest off-state current is mainly attributed to the single halo structure. Fig. 3 shows the potential value in the surface's length of single haloCNTFET and C-CNTFET structures having $V_{gs}=1$ V, $P_{opt}=100$ W/cm² and a wavelength of 456 nm (2.72 eV). Fig. 3 shows that the single halo CNTFET structure has a potential in the length of the channel that has sudden changes. The variation in drain potential does not affect the area of the SH-CNTFET where it is p-doped. That is to say, the drain current, once saturated, is not affected by the drain potential. Drain-induced barrier lowering (DIBL) is reduced significantly.

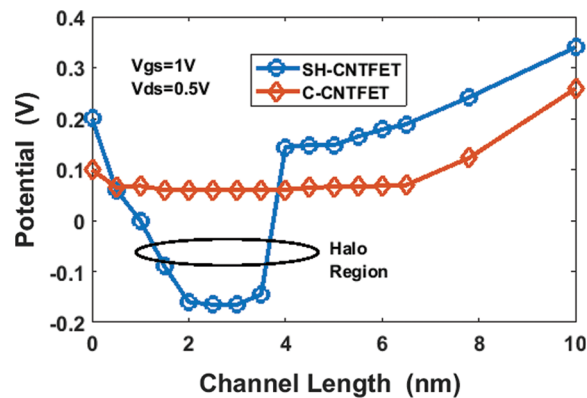


Figure 3: Potential along the surface of the SH-CNTFET structure at $V_{gs}=1$ V, $P_{opt}=100$ W/cm², wavelength (λ) = 456 nm under dark and illumination conditions

Fig. 4 shows the sub-threshold swing of the Single Halo CNTFET and C-CNTFET at $V_{gs}=0.5$ V, $V_{ds}=1$ V, and $P_{opt}=100$ W/cm² in dark and illumination conditions. In CMOS circuits, the standby power dissipation is increased, and the factor contributing to this is the sub-threshold leakage. The theoretical sub-threshold swing is 60 mV/decade at room temperature. This applies to CNTFETs, sub-threshold swings of SH-CNTFET and C-CNTFET at $V_{gs}=0.5$ V and compared in Fig. 4. From Fig. 4, it is evident that when comparing single halo CNTFET with C-CNTFET, the former shows sub-threshold swing. Concerning the smallness of DIBL and the swing of the SH-CNTFET, the proposed SH-CNTFET structure offers better control of Short Channel Effects [25].

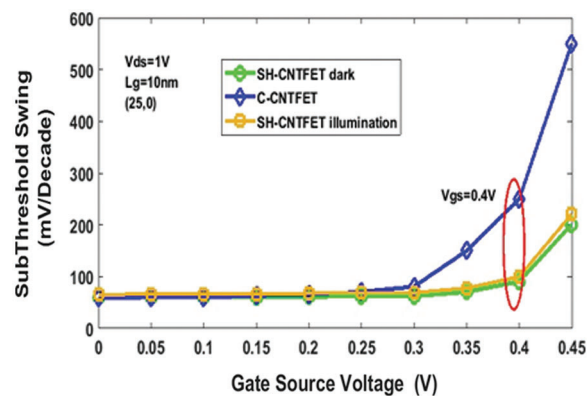


Figure 4: Sub-threshold swing of SH-CNTFET and C-CNTFET with $V_{gs}=0.5$ V, $V_{ds}=1$ V, $P_{opt}=100$ W/cm², wavelength (λ) = 456 nm under dark and illumination conditions

Fig. 5 shows the transconductance characteristics of the SH-CNTFET and C-CNTFET structures at $P_{opt} = 100 \text{ W/cm}^2$, $V_{ds} = 1 \text{ V}$. The simulation is obtained by varying the gate-source voltage from 0 to 1 V, noting and comparing the corresponding current value with C-CNTFET. Then the current is initially there is no increment, and then after 0.5 V, it is increased. The current in illuminated conditions for single halo CNTFET is lower than that for C-CNTFET. This is caused by the increased threshold voltage of the single halo CNTFET and reduced drain conductance. Comparison of the C-CNTFET and SH-CNTFET and the on-off current ratio (I_{on}/I_{off}) results in the observation that the on-off current of the SH-CNTFET is significantly larger. Moreover, the transconductance (g_m) is increased for SH-CNTFETs. Fig. 4 shows that the drain conductance of the SH-CNTFET is reduced [12]. At $V_{gs} = 0.7 \text{ V}$, the transconductance of the SH-CNTFET is $7.8 \times 10^{-5} \text{ mho}$, which is higher than that of the C-CNTFET. The output conductance is $2.8 \times 10^{-6} \text{ S}$.

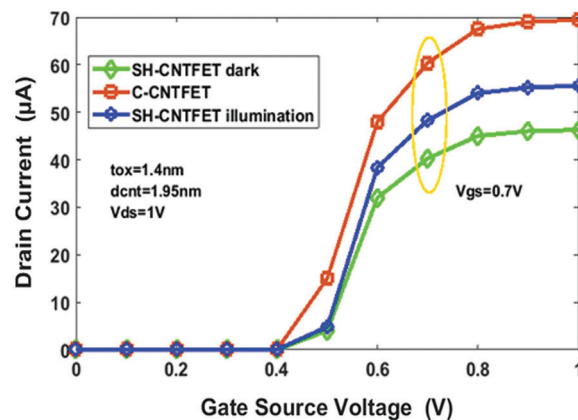


Figure 5: Transconductance of the SH-CNTFET and C-CNTFET with $V_{gs} = 1 \text{ V}$, $V_{ds} = 1 \text{ V}$, $P_{opt} = 100 \text{ W/cm}^2$, wavelength (λ) = 456 nm under dark and illumination conditions

Fig. 6 shows the relationship between the cut-off frequency and the gate-source voltage at 0.8 V. The doping profile of the drain and source of this device achieves the increase of g_m and decrease of channel charge variation vs. gate-source voltage. The gate length of both the C-CNTFET and SH-CNTFET achieves terahertz level cut-off frequencies. A higher cut-off frequency is observed in the halo structure for a fixed gate length. Fig. 5 shows the higher transconductance achieved [12]. As a result, as observed below, there is an increase in the cut-off frequency characteristic. In addition, the cut-off frequency of the SH-CNTFET is 6.45 THz compared to the CNTFET structure, which is 4.3 THz. Fig. 7 shows the gain of SH-CNTFET with various diameters from 0.5 to 2 nm at $P_{opt} = 100 \text{ W/cm}^2$. To achieve a relatively large transconductance, the SH-CNTFET must have a large channel diameter. A minimised hot carrier effect is observed due to a reduction in the peak of the electric field at the drain side. There is a significant increase in the voltage gain [23]. This effect is due to the drastic reduction in the drain conductance of the HD-CNTFET. The larger the transconductance, the greater the gain it will deliver. Consequently, the voltage gain is higher when a nanotube has a larger diameter. The gain of SH-CNTFET is 28.55 db when the diameter of the CNT is set at 1.9 nm. Hence, it can be derived that the lower bandgap is obtained from the (25, 0) chirality semiconductor type CNT. Fig. 8 shows directivity being proportional to the bias voltage, and the directivity of SH-CNTFET at P_{opt} is 100 W/cm^2 . The directivity of SH-CNTFET is $9.4 \text{ cmHz}^{-1}/\text{W}$ at $V_{gs} = 2 \text{ V}$ and $V_{ds} = 1 \text{ V}$. The proposed SH-CNTFET photodetector's efficiency is calculated at $P_{opt} = 100 \text{ W/cm}^2$ and a wavelength of 456 nm. The photons hit the SH-CNTFET photo's surface and produced the photo-excited electron-hole pairs. The responsivity of the proposed SH-CNTFET is 105.48 mA/W, and the efficiency is 75%. Fig. 9 shows the energy band diagram of Au and

CNT. The Au-CNT in the SH-CNTFET device improves the performance of the phototransistor through proper doping.

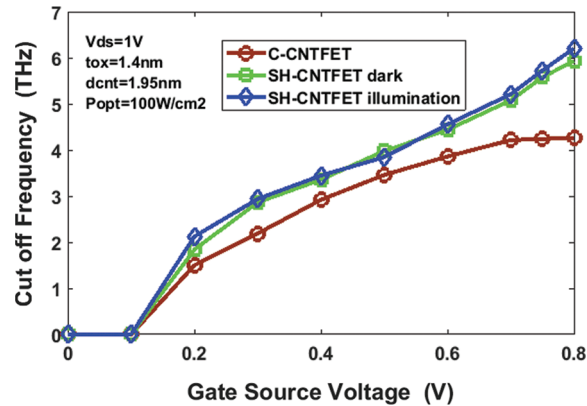


Figure 6: Cut-off Frequency of the SH-CNTFET and C-CNTFET with $V_{gs} = 1$ v, $P_{opt} = 100$ W/cm², wavelength (λ) = 456 nm under dark and illumination conditions

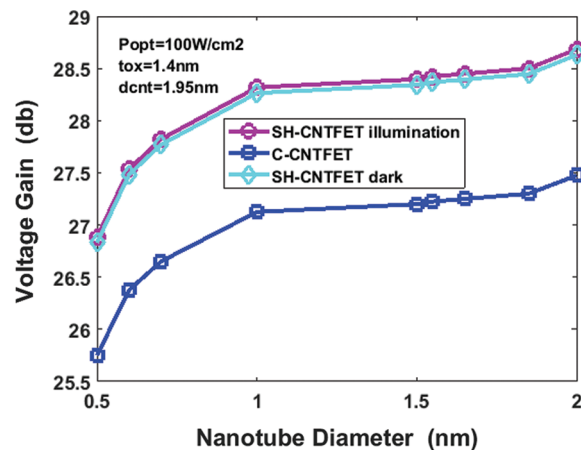


Figure 7: Gain of the SH-CNTFET structure with $V_{ds} = 1$, $P_{opt} = 100$ W/cm², wavelength (λ) = 456 nm under dark and illumination conditions

Table 2 compares the SH-CNTFET with other CNTFET devices at $V_{ds} = 0.5$ V. The sensitivity of the CNT is evaluated at varying visible wavelengths. The light intensity of 456 nm is applied, and the SH-CNTFET's dependence on light intensity is evaluated. The efficiency of the proposed SH-CNTFET devices is observed only in visible light. At the minimum optical incident, with a power of 100 W/cm², it produces better photocurrent when compared with other photo devices such as CNTFET and SWCNT array. When the device is exposed to visible light, photons with higher energy than the CNT's band gap will generate electron-hole pairs and separate the depletion region between metal and CNT to produce photocurrent. In addition, the directivity and responsivity of the SH-CNTFET are significantly increased compared to the SWCNT array. Good sensitivity is observed in the absorption spectrum of the proposed SH-CNTFET and the absorption of visible light.

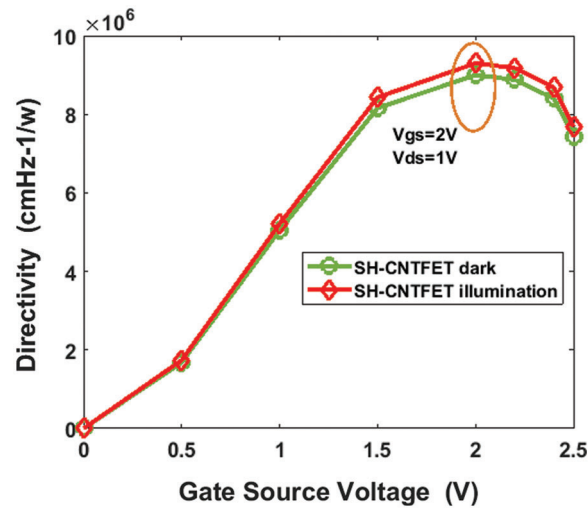


Figure 8: Directivity of SH-CNTFET with $P_{opt} = 100 \text{ W/cm}^2$, wavelength (λ) = 456 nm under dark and illumination

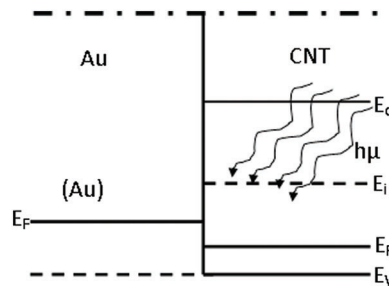


Figure 9: Energy level of Au and CNT

Table 2: A Comparative Output of SH-CNTFET with different CNTFET

Different CNTFET	Optical incident power (Popt)	WaveLength (nm)	Dark condition (Current)	Illumination condition (Current)	Diameter CNT (dcnt)	Directivity (cmHz1/2/ W)	Responsivity (mA/W)	References
SH-CNTEFT	100 W/cm ²	456 nm	20 μA	35 μA	1.95 nm	8.3 × 10 ⁶	105.48	Proposed work
Photo-CNTFET	1.0 mW	633 nm	10 μA	50 μA	0.8 nm	-	121.48	[9]
SWCNT array	1.57 kW	785 nm	10 nA	21 nA	1–2 nm	10.9 × 10 ⁶	65.8	[29]

Table 3 shows the comparison of the SH-CNTFET and DGM-CNFET photodetectors. The diameter of the DMG-CNTFET device is smaller than the proposed SH-CNTFET device. The table shows that increasing the halo doping causes a decrease in conductance. The proposed SH-CNTFET device has these advantages: it provides higher transconductance, cut-off frequency, and gain than the DMG-CNTFET device. $L_g = 10 \text{ nm}$, which is considered to be a fixed undoped channel Length and $L_{shd} = 5 \text{ nm}$, which is the length of

halo doping. The electrical characteristics of SH-CNTFETs are evaluated in terms of current drain characteristics, transconductance, and sub-threshold swing as a function of parameter values.

Table 3: Comparison of DMG-CNTFET and SH-CNTFET photodetector

Parameter/device	DMG-CNTFET photodetector	SH-CNTFET photodetector
Diameter	1.02 nm	1.95 nm
Gate oxide thickness (HfO ₂)(tox)	1.6 nm, k = 16	1.4 nm, k = 30
Doping	Without halo	Halo doping
Dark current	40 μ A	20 μ A
Illumination current	45 μ A	35 μ A
Subthreshold swing	61 mv/decade	61.2 mv/decade
Transconductance	6×10^{-5} mho	7.8×10^{-5} mho
Gain	26.2 db	28.55 db
Cut-off frequency	4.7 THz	6.4 THz
Directivity	8×106	8.3×106

Table 3 compares the DMG-CNTFET and the SH-CNTFET, proving that the dark current is just 20 μ A for the latter method and is better than the previous DMG CNTFET. The SH-CNTFET advances the DMG-CNTFET in sub-threshold swing, transconductance, gain, cut-off frequency and directivity.

5 Conclusion

The SH-CNTFET photodetector proposed in this paper is a single halo-doped with a single gate material of Au, a larger zigzag and smaller energy band gap and a HfO₂ dielectric. The above modifications help achieve reduced leakage current, increased Ion/Ioff ratio, increased cut-off frequency, and increased transconductance and quantum efficiency. The achieved results have also been compared to existing CNTFETs and found to have improved performance. The method used for the characteristic study is found to provide more accurate results. The simulation is achieved through the solution of the Green function method for the SH-CNTFET photodetector under dark and illuminated conditions. The latest technology innovations requiring speed, switching, accuracy and reliability need devices that support such complexity. Also, as the technology is moving into optical innovations, photodetectors such as the proposed device with quantum efficiency are more crucial. In the future, a simulation of the noise analysis of the carbon nanotube FET photodetector can be carried out.

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