

DOI: 10.32604/cmes.2023.023643

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4-Port Octagonal Shaped MIMO Antenna with Low Mutual Coupling for UWB Applications

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Received: 06 May 2022 Accepted: 23 September 2022

ABSTRACT

A 4-port multiple-input multiple-output (MIMO) antenna exhibiting low mutual coupling and UWB performance is developed. The octagonal-shaped four-antenna elements are connected with a 50 Ω microstrip feed line that is arranged rotationally to achieve the orthogonal polarization for improving the MIMO system performance. The antenna has a wideband impedance bandwidth of 7.5 GHz with S₁₁ < -10 dB from (103.44%) 3.5-11 GHz and inter-element isolation higher than 20 dB. Antenna validation is carried out by verifying the simulated and measured results after fabricating the antenna. The results in the form of omnidirectional radiation patterns, peak gain (\geq 4 dBi), and Envelope Correlation Coefficient (ECC) (\leq 0.01) are extracted to validate the suggested antenna performance. As well, time-domain analysis was investigated to demonstrate the operation of the suggested antenna in wideband applications. Finally, the simulated and experimental outcomes have almost similar tendencies making the antenna suitable for its use in UWB MIMO applications.

KEYWORDS

MIMO antenna; low mutual coupling; time domain analysis; UWB antenna

1 Introduction

Recently, the rapid evolution of modern communication systems motivates researchers to steer their work to wideband technology. The Federal Communications Commission (FCC) is approved bandwidth extended from 3.1 to 10.6 GHz (7.5 GHz) to be utilized in UWB systems [1]. Antennas are considered the most vital part of the UWB modern communication systems because they should fulfill several demands such as wideband feature, compact size, low price, omnidirectional patterns, and the simplicity of integration with the UWB system [2–6].



Multiple-input- multiple-output (MIMO) is considered another technology utilized in communication systems. In MIMO systems, several antennas are used in the transmitting and receiving ends to enhance the data rate and signal quality without increasing the bandwidth and the power of the transmission. The UWB technology is suitable for application scenarios where communication needs to be carried out at a smaller distance because of its weak power and the multipath signal reception problem. So, MIMO technology should be integrated with the UWB technology to increase the channel capacity and fix the multipath signal reception problem. In MIMO systems, the integration of the compact MIMO antennas with the other components is considered the main issue because of the limited available space for the antennas. So, adding antenna elements close to each other is considered the best choice. However, the mutual coupling between elements is increased and isolation techniques should be utilized to reduce the effect of the mutual coupling which increases the complexity of the antenna design [7-11].

Several procedures have been studied for isolation improvement among the antenna elements in the UWB MIMO system. Stubs with different forms are used to enhance the isolation between ports, as discussed in [11–15]. While in [16–20], the isolation is improved by placing the antenna elements in a perpendicular arrangement without any complex use of decoupling structure. Reference [11] discussed four elements of the MIMO antenna using stubs for reducing the undesired mutual coupling. A floating parasitic isolation structure to reduce the mutual coupling is placed between antenna elements [12]. A rectangular stub is added between two antennas, as investigated in [13]. In [14], a resonating stub connected in the ground leads to reduced mutual coupling among the ports. Grounded stubs are used in [15] to improve the isolation between four elements UWB MIMO antenna. In [16], a four ports UWB monopole antenna with a slotted ring radiator and placed orthogonally without utilizing a decoupling structure is introduced. Four elements UWB antenna with CPW fed and four circular slots in the circular radiator are placed orthogonally to enhance the isolation between elements is investigated in [17]. In [19,20], a four ports UWB antenna placed orthogonally to reduce the mutual coupling between elements and without using a decoupling structure is discussed and investigated.

For more simplicity, this paper suggested a 4-port UWB MIMO antenna without using decoupling structures. A 50 Ω microstrip feed line is utilized to feed the four octagonal shape radiator antennas. The isolation is enhanced by adding the elements in a perpendicular arrangement to achieve isolation higher than 20 dB between elements for the desired band of 3.5–11 GHz. The simulation of the antenna is carried out using CST studio. The antenna has a size of $62.5 \times 60.5 \times 1.6 \text{ mm}^3$. The antenna has omnidirectional patterns with a peak gain of >4 dB. The results of the diversity performance of the MIMO are ECC < 0.01, DG \geq 9.95 within the operating bands. Finally, from the tested results in the form of S-parameters and radiation patterns, it can be claimed that the suggested antenna can be operated in UWB communications.

2 UWB Single Antenna

The development process of the UWB antenna by changing the radiator shape and ground length to match the UWB technology requirement, such as wide bandwidth, small size, and simple design is shown in Fig. 1.

The S_{11} simulated results are carried out to obtain the proposed structure, as shown in Fig. 2. A rectangular patch (antenna 1) is started as an initial design with a partial ground plane of 10.5 mm as shown in Fig. 1a. The simulated S_{11} of antenna 1 shows that the antenna operated from (66.66%) 3–6 GHz as illustrated in Fig. 2. To improve the antenna bandwidth, an octagonal shape radiator is utilized as depicted in Fig. 1b (antenna 2) with the same length as the ground plane. Antenna 2 is

operated from (78.26%) 3.5–8 GHz with S_{11} lower than -10 dB and from 10–11 GHz as shown in Fig. 2. Finally, to improve the band from 8–10 GHz small rectangular slot is etched in the bottom as shown in Fig. 1c (antenna 3). The antenna has an impedance bandwidth spanning from (103.44%) 3.5–11 GHz with S_{11} lower than -10 dB, as seen in Fig. 2, which is suitable for the UWB technology.



Figure 1: UWB single antenna development (a) Antenna 1 (b) Antenna 2 (c) Proposed antenna



Figure 2: The S_{11} results of the three antenna structures

To design the antenna to operate at the resonant frequency of 3.5 GHz, the value of the radiator radius is calculated based on the cavity model to excite the dominant mode TM_{110} as Eq. (1) [21]. Where, f_r , ε_r , and h are the resonant frequency, dielectric constant, and the thickness of the substrate,

respectively. Also, to increase the operating bandwidth, the partial ground plane with a rectangular slot is employed as shown in Fig. 1c.

$$R = \frac{F}{\left(1 + \frac{2h}{\pi F \varepsilon_r} \left(In \left(\frac{\pi F}{2h}\right) + 1.7726 \right) \right)^{0.5}}, F = \frac{8.791 \times 10^9}{f_r \sqrt{\varepsilon_r}}$$
(1)

To show the effect of the antenna structure parameters such as the radius (R), the ground length (Lg), the ground slot width (Ws), and length (Ls) on the antenna performance, parametric study analyses are carried out as shown in Fig. 3. As illustrated in Fig. 3a, when the radius (R) is increased from 7 to 11 mm, the first band is shifted from 4 to 3 GHz based on the Eq. (1). The radius (R) equal to 9 mm is used to achieve the recommended results. As shown in Fig. 3b, when the ground length (Lg) is increased from 8.5 to 10.5 mm, the antenna bandwidth and matching are affected. The ground length (Ws) is increased from 3 to 7 mm, the antenna matching impedance is affected. The ground slot width (Ws) equal to 5 mm is used to achieve the proposed results. Finally, when the ground slot length (Ls) is increased from 1 to 5 mm, the antenna matching impedance is affected. The ground slot length (Ls) equal to 3 mm is used to achieve the proposed results.

Based on the previous parametric study, the proposed 2D configuration (top and back) of the UWB antenna structure is presented in Fig. 4a, and its fabricated prototype is illustrated in Fig. 4b. The antenna has an octagonal shape radiator connected with a microstrip line matched at 50 Ω . A defected arrangement on the ground plane and a small rectangular slot etched on it are utilized to achieve the required wideband as discussed in the previous paragraph. A cost-efficient FR4 substrate is used in the fabrication process where thickness is 1.6 mm, tan δ is 0.003, and $\varepsilon r = 4.4$. The antenna is checked and tested using four ports vector network analyzer (VNA) R&S ZVB 20.

The S_{11} results are illustrated in Fig. 5. It is clear from the simulated results that, the antenna works in the frequency band from 3.5–11 GHz with $S_{11} \le -10 \text{ dB}$ while the measured results show almost comparable bandwidth ranging from (97.29%) 3.8–11 GHz However, a small shift between results can be noticed; this could be due to the fabrication and soldering procedure.



Figure 3: (Continued)



Figure 3: The parametric study results of the proposed octagonal shape antenna (a) The effect of the radius (R) (b) The effect of the ground length (Lg) (c) The effect of the ground slot width (Ws) (d) The effect of the ground slot length (Ls)



Figure 4: Single antenna geometry (a) 2D configuration layout (b) Fabricated prototype photo



Figure 5: Simulated and measured S₁₁ results of UWB single antenna

3 Proposed 4 Ports Antenna

To convert the single port to the four ports, the two ports with two configurations side by side and orthogonal orientations with the same edge-to-edge distance of 9 mm between elements are discussed as illustrated in Fig. 6. The orthogonal orientation has good results than the side-by-side configuration. The mutual coupling is enhanced by using the orthogonal orientation without adding extra structures. So, the orthogonal orientation is utilized in the four ports antenna.



Figure 6: The effect of the two elements' orientation on the UWB antenna performance at port (a) S_{11} (b) Isolation (S_{21})

The single and the two ports UWB antenna discussed in the previous section is arranged to achieve and realize the proposed 4 ports MIMO antenna with $62.5 \text{ mm} \times 60.5 \text{ mm}$ as shown in Fig. 7. The antenna in port 1 has an orthogonal orientation with the antenna on port 2, and port 4, and has the same orientation but in opposite direction to the antenna in port 3. This arrangement is utilized to produce the polarization diversity which in turn increases the isolation between elements. The fabricated prototype is demonstrated in Fig. 7b where the same substrate of the single antenna is used in the fabrication process. The measured results are extracted using four ports R&S ZVB 20 VNA from port 1 because of the symmetry property and the other three ports have the same results.



Figure 7: 4-ports UWB MIMO antenna (a) 2-D configuration (b) Fabricated prototype photo

Fig. 8 illustrates the proposed MIMO UWB antenna results at port 1. It is noticed that the simulated results are operated from (103.44%) 3.5–11 GHz with $S_{11} \leq -10 \text{ dB}$ while the measured outcomes show that the antenna almost matched the simulated impedance bandwidth spanning from (97.29%) 3.8–11 GHz. The same trend is observed in both the results with a minor shift due to the fabrication and SMA connector losses. The simulated and measured isolation results at port 1 (S_{21} , S_{31} , S_{41}) are shown in Fig. 9. It is seen that the measured S_{21} and S_{41} (perpendicular orientation) have values lower than -20 dB while the S_{31} has a value lower than -18 dB. Also, the results have good matching between them with the small shift as shown in Fig. 9.

The simulated current distributions of the UWB MIMO antenna at 5 and 8 GHz at four ports are shown in Fig. 10. The current is concentrated around the octagonal patch proving that the antenna radiated at these two frequency bands. Also, the current is confined around the antenna with a very small amount of current dissipated to the other ports which confirm the high isolation between ports.

The 3 D radiation pattern simulation results at 4.5 and 9.5 GHz are illustrated in Fig. 11 to validate the diversity of the patterns. As shown in Fig. 11, when the suggested port is excited the remaining three

ports are connected to 50Ω load. Also, from Fig. 11 it is seen that the 90° phase difference between the ports is achieved to confirm the polarization diversity which in turn enhances the MIMO system.

The proposed MIMO antenna radiation patterns and peak gain is measured. Fig. 12 depicts the setup inside an anechoic chamber.



Figure 8: Simulated and measured S₁₁ results of the proposed 4 ports UWB antenna at port 1



Figure 9: (Continued)



Figure 9: Isolation results of the 4 ports UWB MIMO antenna at port 1 (a) S_{21} (b) S_{31} (c) S_{41}



Figure 10: Simulated results of current distributions of the proposed 4 ports UWB antenna (a) @ 5 GHz (b) @ 8 GHz



Figure 11: The 3D radiation pattern results of the proposed 4 ports UWB MIMO antenna (a) 4.5 GHz (b) 9.5 GHz



Figure 12: Radiation pattern measurements set up of the 4 ports UWB MIMO antenna @ port 1

The antenna is measured at port 1 where the additional ports are terminated using 50 Ω load. The normalized radiation patterns along x-z/y-z planes at 4.5, 6.5, and 9.5 GHz are illustrated in Fig. 13. It is seen that the antenna has an omnidirectional pattern in the x-z plane and semi-bidirectional in the y-z plane. However, at the higher frequency band, the antenna has omnidirectional patterns in both planes.



Figure 13: Normalized radiation pattern results of the proposed 4 ports UWB MIMO antenna @ port 1 (a) 4.5 GHz (b) 6.5 GHz (c) 9.5 GHz

Fig. 14 shows the peak gain and simulated total and radiation efficiencies of the MIMO antenna at port 1. From Fig. 14a, the results show that the antenna has peak gain extended from 3.5 to 5.5 dBi within the operated frequency bands with good agreements between the simulated and experimental results. While from Fig. 14b, the results show that the antenna has radiation efficiency extended from 77% to 95% and total efficiency ranging from 70% to 90% within the operated frequency bands. The reduction of the total efficiency especially at higher frequency band is due to the FR4 lossy substrate and the matching impedance of the antenna up to 7 GHz as shown in Fig. 8.



Figure 14: The proposed 4 ports UWB antenna @ port1 (a) Peak gain results (b) Efficiency

The ECC and DG are two parameters to measure the MIMO performance. The ECC measures the relationship between antenna elements in the MIMO system and its value is recommended to be lower than 0.5 and it can be calculated from Eq. (2) [22–24].

$$\rho_e = \frac{\left| \int \int_{4\Pi} \left[s_i(\theta,\varphi) \right] * \left[s_i(\theta,\varphi) \right] d\Omega \right] \right|}{\int \int_{4\Pi} \left| s_i(\theta,\varphi) \right|^2 d\Omega \int \int_{4\Pi} \left| s_i(\theta,\varphi) \right|^2 d\Omega}$$
(2)

Fig. 15 shows the ECC at port 1. It is noticed that the ECC > 0.01 within the band of interest. As well as the DG is connected with the ECC and is calculated from Eq. (3) [22–24].

$$DG = 10 \times \sqrt{1 - |ECC|}$$
(3)



Figure 15: ECC results of the proposed 4 ports UWB antenna

Fig. 16 illustrates the DG where for the band of interest, the value of 9.95 dB is achieved. Also from the results in Figs. 15 and 16, it is noticed that the proposed antenna has good MIMO results which make our antenna suitable for MIMO diversity systems for UWB communication applications.



Figure 16: DG results of the proposed 4 ports UWB antenna

4 Simulated Time Domain Analysis

The proposed antenna is checked using time-domain analysis to validate its performance. The antenna in UWB technology transmits a narrow band pulse which means this pulse has a wide-band response in the frequency domain. So, time-domain features as a function of transmission coefficient S_{21} , S_{21} phase, and group delay are important parameters to show the ability of the antenna to be operated in UWB applications. Fig. 17 illustrates the simulated time-domain setup at different three layouts (face to face, face to side, and side by side). The two similar antennas are placed in three different configurations maintaining a distance of 210 mm (2.45 times λ_0 at 3.5 GHz) between them. One antenna behaves as the transmitter (Tx) while the other acts as the receiver (Rx).



Figure 17: Simulated time-domain analysis setup at different layout

Fig. 18 shows the magnitudes of the S_{21} , S_{21} phases, and the unwrapped S_{21} phase. It is clear from Fig. 18a that the three antenna configurations have an S_{21} value lower than -30 dB throughout the entire frequency band with a stable value from 3.5-9 GHz while at the higher frequency band from 10 to 11 GHz the value is reduced lower than -40 dB, especially at the face to face and face to side configurations. As well, the S_{21} phase is shown in Fig. 18b and the unwrapped phase for better observation is illustrated in Fig. 18c to know the linearity features within the wideband operation. It is seen from the phase curves that the antenna has linear operation within the frequency band at different configurations. For more confirmation, group delay results are illustrated in Fig. 19. It is seen that the group delay has a stable value of around 1.1 ns within the operating frequency band with around 0.1 ns variation at the end of the band, especially at the two face-to-face and face-toside configurations. Table 1 illustrates the suggested work in comparison to other former works. The suggested work has a simple design, high isolation, and high gain which suggests it be employed in the UWB MIMO systems.



Figure 18: Simulated results at the different layout (a) S_{21} (b) S_{21} phase (c) S_{21} unwrapped phase



Figure 19: Simulated group delay results at different layout

Ref.	Size (mm ²)	εr/thickness (mm)	B.W GHz	No. of antennas	Gain (dBi)/ Efficiency (%)	Isolation (dB)	ECC/DG
[6]	40×30	4.4/1.6	3.20-5.85	4	3.5/85	>17.5	<0.15/9.9
[11]	42×42	4.4/1.6	3.2–12	4	4/80	≥17	0.01/9.96
[12]	45.5×33	3.55/1.5	3-12	2	4/75–90	≥ 20	0.09/-
[13]	66×36	3.38/0.813	2.6-12.5	2	4/-	≥ 20	0.001/9.97
[15]	45×45	4.4/1.6	2-10.6	4	3/-	≥17	0.005/-
[16]	38×38	3.2/0.762	3–15	4	3.5/65–79	≥ 20	<0.5/-
[17]	81×87	4.4/1.6	3.03-10.74	4	5/60-87	≥20	0.1/-
[19]	48×52	4.4/1.6	2.7 - 11	4	3.5/80	>20	0.0004/9.85
[20]	26×55	4.4/1.6	3.1-12.3	4	2.5/-	>20	<0.5/-
This work	62.5×60.5	4.4/1.6	3.5–11	4	4/70–90	≥20	< 0.01/9.95

Table 1: Comparison between our suggested work and other selected designs

The antenna is compared with the existing MIMO antennas from the literature in Table 1. It is observed that the antenna has a low profile, wide bandwidth, 4-port configuration, high gain, low mutual coupling without using a complex decoupling structure, and low ECC value that qualifies the antenna for its use in UWB MIMO applications.

5 Conclusion

A four-port MIMO antenna for UWB operation has been investigated in this paper. The suggested antenna element, as the initial design, is first fabricated and tested. After that, 4 ports MIMO antenna with perpendicular arrangement has been discussed, simulated, fabricated, and tested using VNA. As well, radiation patterns, peak gain, and MIMO performance have been investigated. The suggested antenna has been worked at a frequency band extending from 3.5 to 11 GHz with more than 20 dB,

4 dBi isolation between ports, and peak gain, respectively. The time response has been discussed to validate the suggested antenna to operate in wideband applications. Finally, the proposed antenna has been compared with other antennas, and the results show that the suggested antenna can be used in UWB communication.

Funding Statement: The authors received no specific funding for this study.

Conflicts of Interest: The authors declare that they have no conflicts of interest to report regarding the present study.

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